

4/21/23

# RECITATION 9B FIR FILTER DESIGN

## FIR DESIGN TECHNIQUES

↳ DESIGN USING WINDOWS

↳ FREQUENCY SAMPLING DESIGN

↳ PARS- M. CLEGGAN ALGORITHM

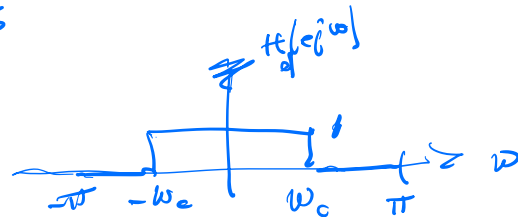
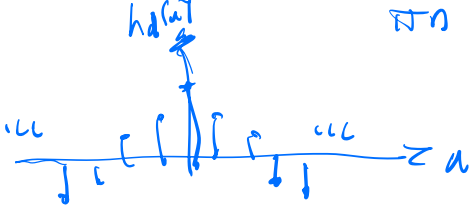
$$h[n] = h_d\left(n - \frac{M}{2}\right) \left\{ w[n] \right\}$$

↑
↑
↑

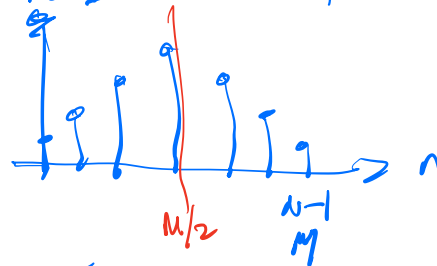
FIR  
LINEAR  
PHASE
 ↑
 SCALED  
IDEAL  
RESPONSE  
(ZERO PHASE,  
INFINITE)
 ↑
 FINITE  
WINDOW,  
SYMMETRIC  
ABOUT  $M/2$

## PROTOTYPE (IDEAL) FILTERS

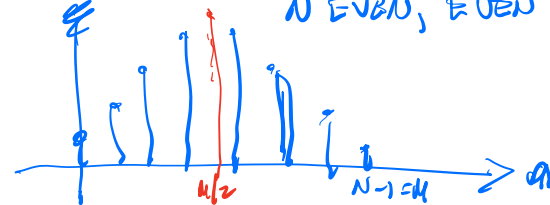
LPF  $h_d[n] = \frac{\sin(\omega_c n)}{\pi n}$ ,  $\omega_c$



TYPE I  $N$  ODD, EVEN SYMM.



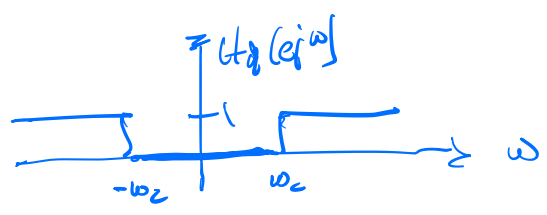
TYPE II  $N$  EVEN, EVEN SYMM.



HPF

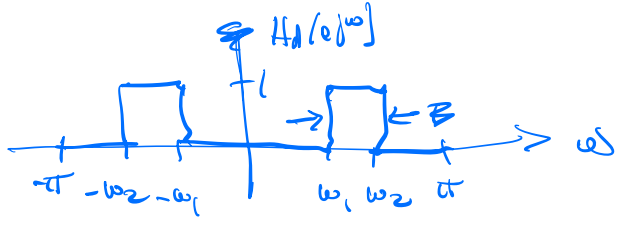
$$h_g[n] = \delta[n] - \frac{\sin(\omega_c n)}{\pi n}$$

TYPE I ONLY



BPB

$$h_d[n] = \frac{\sin(\omega_2 n)}{\pi n} - \frac{\sin(\omega_1 n)}{\pi n}$$



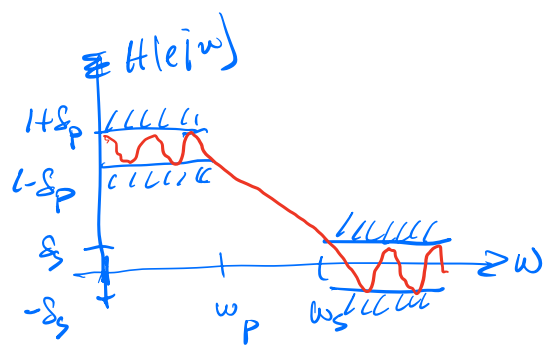
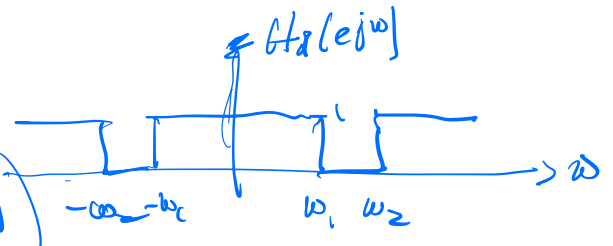
let  $\omega_c = \frac{\omega_2 + \omega_1}{2}$ ,  $B = \omega_2 - \omega_1$

$$h_d[n] = \frac{2 \sin\left(\frac{B}{2} n\right) \cos(\omega_c n)}{\pi n}$$


BAND STOP

$$h_b[n] = \delta[n] - \left( \frac{\sin(\omega_2 n)}{\pi n} - \frac{\sin(\omega_1 n)}{\pi n} \right)$$

TYPE I ONLY



## WINDOW DESIGN

$$h[n] = h_d[n - \frac{M}{2}] w[n] \Leftrightarrow H(e^{j\omega}) = H_d(e^{j\omega}) \cdot e^{j\omega \frac{M}{2}} \cdot \frac{1}{2\pi} W(e^{j\omega})$$


## PROCEDURE

1. CHOOSE WINDOW SHAPE BASED ON DESIRED STOP BAND ATTENUATION
2. CHOOSE WINDOW LENGTH BASED ON TRANSITION BAND WIDTH

Ex.  $\omega_p = 0.25\pi$   
 $\omega_s = 0.3\pi$  }  $\Delta\omega = 0.05\pi$

PASSBAND RIPPLE =  $\pm 1$  dB

STOPBAND RIPPLE  $< -60$  dB

$20 \log_{10}(1 - \delta_p) < -1$

$20 \log_{10}(\delta_s) < -60$

60 dB STOPBAND ATTENUATION  $\rightarrow$  BLACKMAN WINDOW

$$\Delta\omega = 0.05\pi = \frac{12\pi}{M}$$

$$M = \frac{12}{0.05} = 240, \quad \omega_c = \frac{\omega_p + \omega_s}{2} = 0.275\pi$$

$$h[n] = h_d[n - \frac{M}{2}] \cdot w[n]$$

$$= \frac{\sin(0.275\pi(n - 140))}{\pi(n - 140)} \cdot \left( 0.42 - 0.5 \cos\left(\frac{2\pi n}{240}\right) + 0.08 \cos\left(\frac{4\pi n}{240}\right) \right)$$

## Kaiser Window Design

$$w(n) = \frac{I_0 \left[ \beta \left( 1 - \left( \frac{n-d}{2} \right)^2 \right)^{1/2} \right]}{I_0(\beta)}$$

$$0 \leq n \leq M$$

$$0, \text{ ELSE}$$

$\beta$  SHAPE PARAM.  
 $d = M/2$

### DESIGN PROCEDURE

$$\Delta W = W_s - W_p = 0.05\pi$$

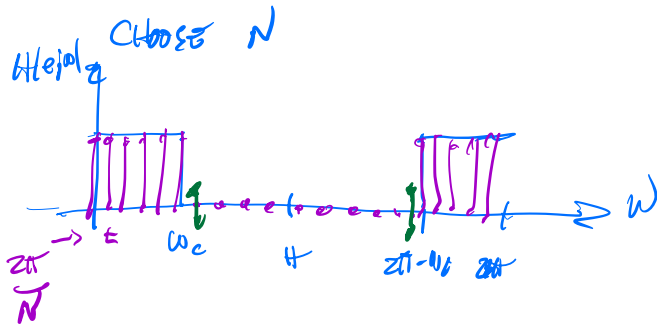
$$A = -20 \log_{10}(\delta_s) = 60$$

$$1. \quad \beta = \begin{cases} 0.1102 (A - 8.7), & A \leq 50 \\ 0.5842 (A - 21)^4 + 0.07886 (A - 21), & 21 \leq A \leq 50 \\ 0, & A \leq 21 \end{cases}$$

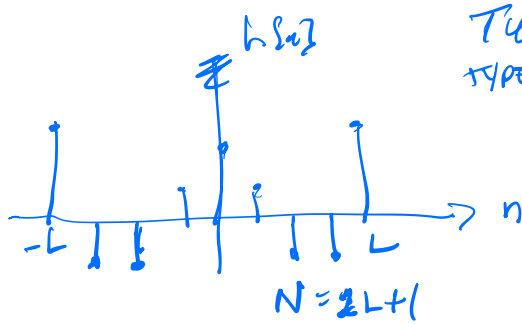
$$\beta = (0.1102)(51.3) = 5.6533$$

$$2. \quad M = \frac{A - 8}{(2.285)(\Delta W)} = 144.9 \Rightarrow 145$$

# FREQUENCY - SAMPLING DESIGN



## PARIS-Mc CULLAN ALGORITHM



TYPE I    OQPA    Z.Z.1  
 TYPE II    OSYA    Z.Z.2

$$H(e^{j\omega}) = \sum_{n=-L}^L h[n] e^{-j\omega n}$$

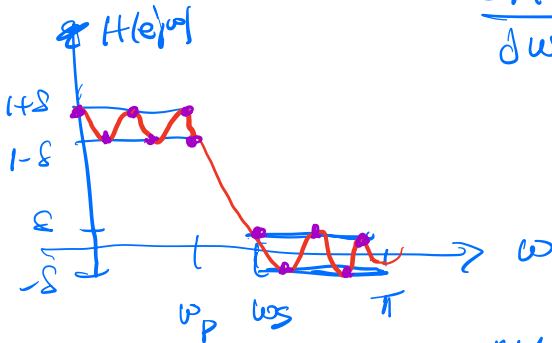
$$= h[0] + \sum_{n=1}^L h[n] 2 \cos(\omega n)$$

ALTERNATION THM.  
 FOR A TYPE I TO BE  
 OPTIMAL, NEED AT LEAST

$$H(e^{j\omega}) = \sum_{l=0}^L b_l (\cos(\omega))^l$$

$B=0$

L+2 ALTERNATIONS



$$\frac{dH(e^{j\omega})}{d\omega} = -n \sum_{l=0}^{L-1} b_l (\cos(\omega))^{l-1} \sin(\omega n)$$

$l=0$      $l-1$      $0$  FOR  $\omega=0, \pi$

ZEROS FOR  $0 < \omega < \pi$

MAX ACTS WOULD BE

$$L-1 + 2 + 2 = L+3$$

$\omega_p, \omega_s$      $\neq$      $0, \pi$

# PARKS - McCLELLAN ALGORITHM

GIVEN  $L, \omega_p, \omega_s,$

$\delta_p / \delta_s$

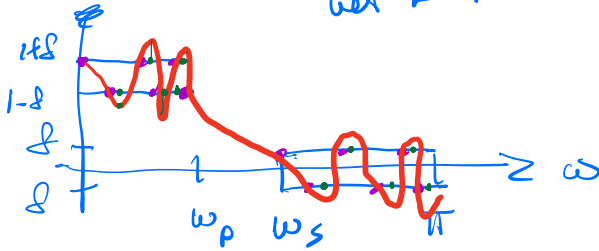
GUESS

1. CHOOSE  $L+2$  FREQS.

INCLUDE  $\omega_p, \omega_s, 0$  AND

$L-1$  OTHERS

Let  $L=10$



2. FITTED CHEBYSHEV POLYNOMIAL THROUGH  
ALL POINTS

3. Remez EXCHANGE: REPLACE ALL FREQS  
BY OBSERVED ALL FREQS.