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FFT ALTERNATE STRUCTURES

[OSPP 9.2-9.5]

$$\text{DFT: } X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] W_N^{-nk}$$

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}$$

FOR COMPUTING IDFT

FLIP SIGNS ON W_N^{nk}

ALT. METHOD FOR IDFT:

REGULAR DFT $X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}$

* COMPLEX CONJUGATE INPUT $X^*[k] = \sum_{n=0}^{N-1} x^*[n] W_N^{nk}$

* COMPLEX CONJUGATE RESULT,

DIVIDE BY N

$$\frac{1}{N} \left(\sum_{n=0}^{N-1} x^*[n] W_N^{nk} \right)^*$$

$$\left(e^{j \frac{2\pi nk}{N}} \right)^* = e^{-j \frac{2\pi nk}{N}} = W_N^{-nk}$$

$$X^*[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] W_N^{-nk}$$

TO COMPUTE IDFT of $X[k]$ 1. COMPUTE $X^*[k]$ 2. PERFORM FORWARD DFT ON $X^*[k]$

3. COMPLEX CONJUGATE RESULT, DIVIDE BY N

DECIMATION IN FREQUENCY

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}$$

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} x[n] W_N^{nk} + \sum_{n=\frac{N}{2}}^{N-1} x[n] W_N^{nk}$$

let $l = n - \frac{N}{2}$
 $n = l + \frac{N}{2}$

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} x[n] W_N^{nk} + \sum_{l=0}^{\frac{N}{2}-1} x[l + \frac{N}{2}] W_N^{(l + \frac{N}{2})k}$$

$$W_N^{(l + \frac{N}{2})k} = W_N^{lk} \cdot W_N^{\frac{N}{2}k}$$

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} x[n] W_N^{nk} + (-1)^k \sum_{n=0}^{\frac{N}{2}-1} x[n + \frac{N}{2}] W_N^{nk} = W_N^{ek} (-1)^k$$

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} (x[n] + (-1)^k x[n + \frac{N}{2}]) W_N^{nk}$$

For k EVEN, let $k = 2r$

k EVEN

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} (x[n] + x[n + \frac{N}{2}]) W_N^{2kn} W_{N/2}^{kn}$$

$\frac{N}{2}$ -PT. DFT of $(x[n] + x[n + \frac{N}{2}])$

For k ODD, let $k = 2r + 1$

k ODD

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} (x[n] - x[n + \frac{N}{2}]) W_N^{n(2r+1)} W_{N/2}^{rn}$$

$$W_N^{n(2r+1)} = W_N^{rn} \cdot W_N^{rn} = W_{N/2}^{rn}$$

$$X[k] = \sum_{n=0}^{\frac{N}{2}-1} (x[n] - x[n + \frac{N}{2}]) W_{N/2}^{rn} W_{N/2}^{rn}$$

$\frac{N}{2}$ -PT. DFT of $(x[n] - x[n + \frac{N}{2}]) W_{N/2}^{rn}$

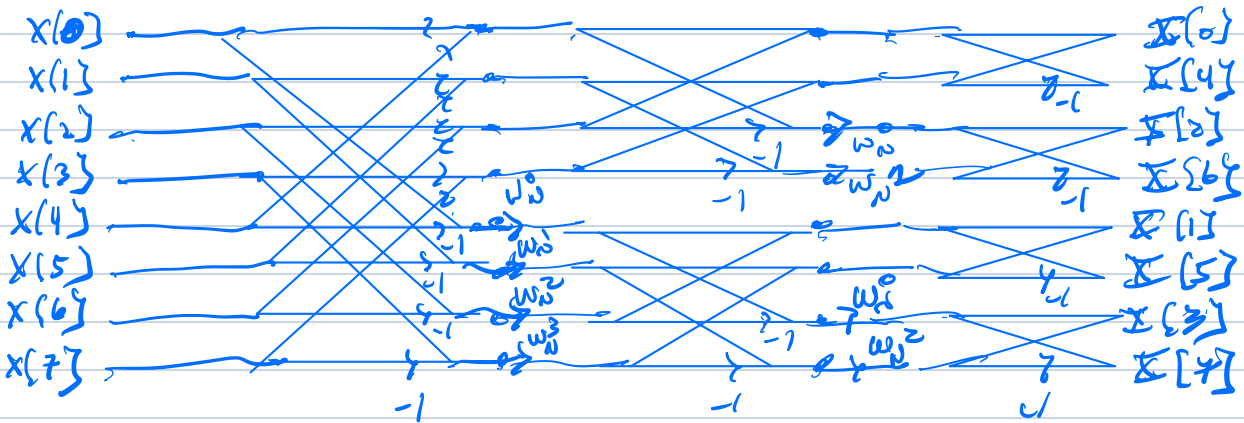
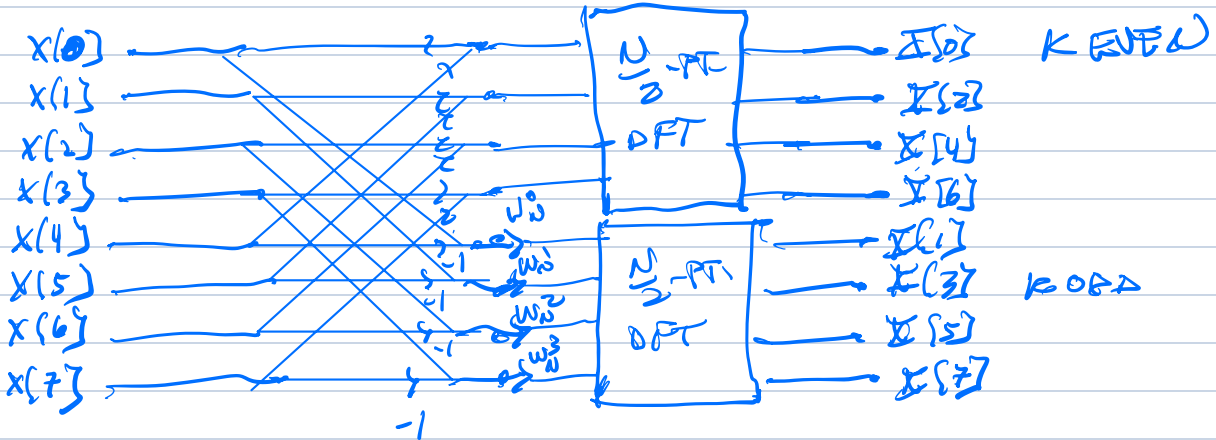
k EVEN

$$\frac{N}{2} - \text{PT. DFT of } [x(n) + x(n + \frac{N}{2})]$$

k ODD

$$\frac{N}{2} - \text{PT. DFT of } (x(n) - x(n + \frac{N}{2})) w_N^n$$

$N=8$



TRANSPOSES IN SIGNAL FLOW DIAGRAM

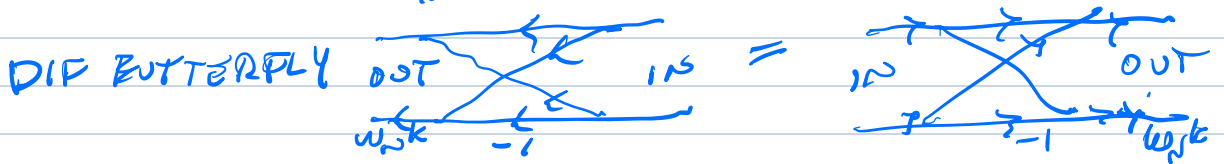
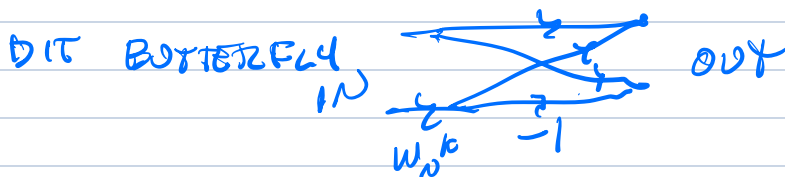
MATRIX TRANSPOSITION THEOREM (S.I. MASON)

1. GIVEN SIGNAL FLOW GRAPH

2. REVERSE ARROWS

3. INTERCHANGE INPUT & OUTPUT

⇒ NETWORK RESPONSE IS THE SAME!



DECIMATION IN TIME FOR $N \neq 2^L$

COMPOSITE $N = p_1 p_2 p_3 \dots p_L$

let $q_1 = \frac{N}{p_1}$, $q_2 = \frac{N}{p_1 p_2}$, $q_3 = \frac{N}{p_1 p_2 p_3}$...

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}$$

let $N = p_1 p_2$

$$X[k] = \sum_{r=0}^{q_1-1} x[p_1 r] W_N^{p_1 r k} + \sum_{r=0}^{q_1-1} x[p_1 r + 1] W_N^{(p_1 r + 1)k} + \dots + \sum_{r=0}^{q_1-1} x[p_1 r + (p_1 - 1)] W_N^{(p_1 r + (p_1 - 1))k}$$

CONSIDER l^{th} SUM: $\sum_{r=0}^{q_1-1} x[p_1 r + l] W_N^{(p_1 r + l)k}$

$$= W_N^{p_1 r k} \cdot W_N^{lk} = W_{q_1}^{rk} \cdot W_N^{lk}$$

$$X[k] = \sum_{b=0}^{p_1-1} W_N^{bk} \sum_{r=0}^{q_1-1} x[p_1 r + b] W_{q_1}^{rk}$$

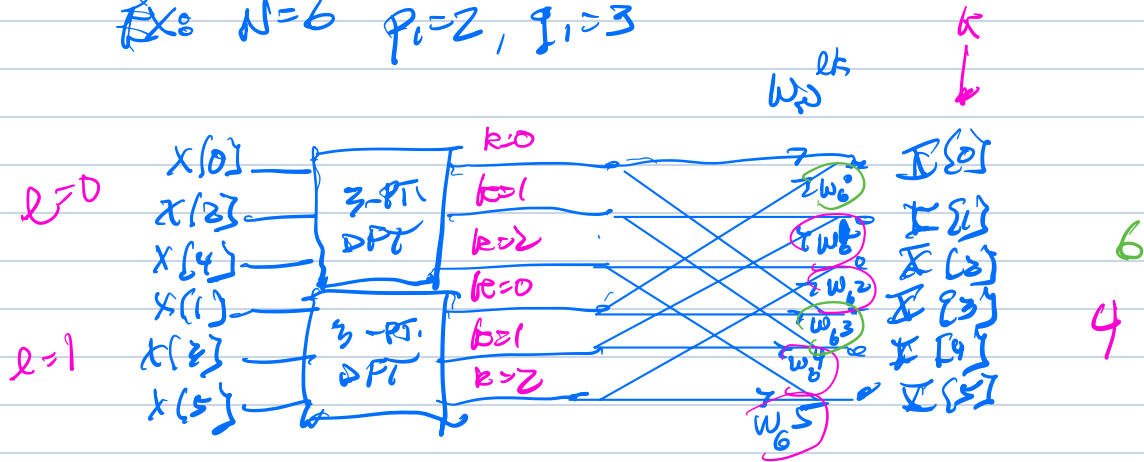
q_1 -PT-DFT of $x[p_1 r + b]$

EX: $N=6$, $p_1=2$, $q_1=3$

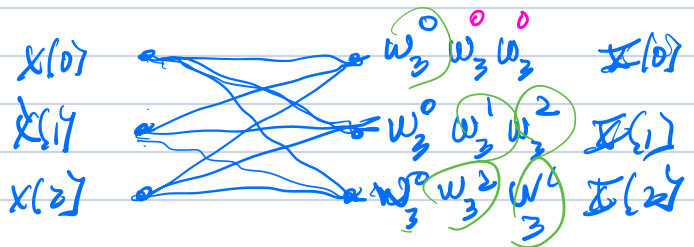
$$X[k] = \sum_{n=0}^{P_i-1} W_N^{nk} \sum_{r=0}^{Q_i-1} x[p_i r + n] W_{Q_i}^{rk}$$

2ⁱ-pt. DFT of $x[p_i r + n]$

Ex: $N=6$ $P_i=2$, $Q_i=3$



6
4



$6^2 = 36$ MULTIPLY

4 + 4 12 NOW STAY MULTIPLY

ATU: $6 + 2(9)$ 29 MULTI ALL TOGETHER