

**18-491/691 Lecture #16**  
**FAST FOURIER TRANSFORM**  
**ALTERNATE IMPLEMENTATIONS**

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# Introduction

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- In our lecture last Monday we described and discussed the basic decimation-in-time Cooley-Tuckey fast Fourier transform algorithm for DFT sizes that are integer powers of 2 (radix 2)
- Today we will discuss some variations and extensions of the basic FFT algorithm:
  - Computation of the inverse FFT
  - One further trivial efficiency
  - Alternate forms of the FFT structure
  - The decimation-in-frequency FFT algorithm
  - FFT structures for DFT sizes that are not an integer power of 2

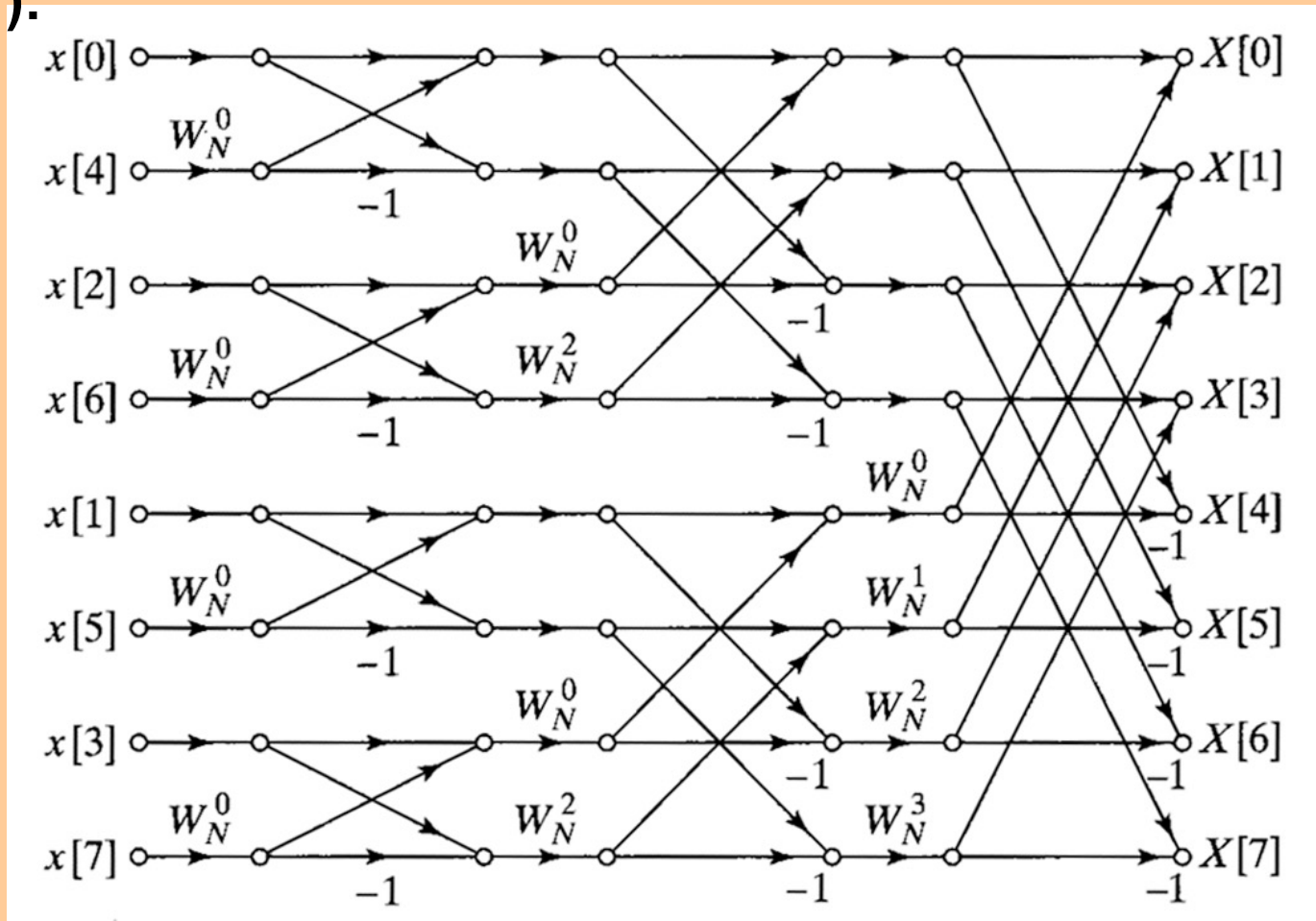
# Alternate FFT structures

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- We developed the basic decimation-in-time (DIT) FFT structure in the last lecture, but other forms are possible simply by rearranging the branches of the signal flowgraph
- Some issues to consider:
  - Natural or bit-reversed input and output?
  - In-place computation?

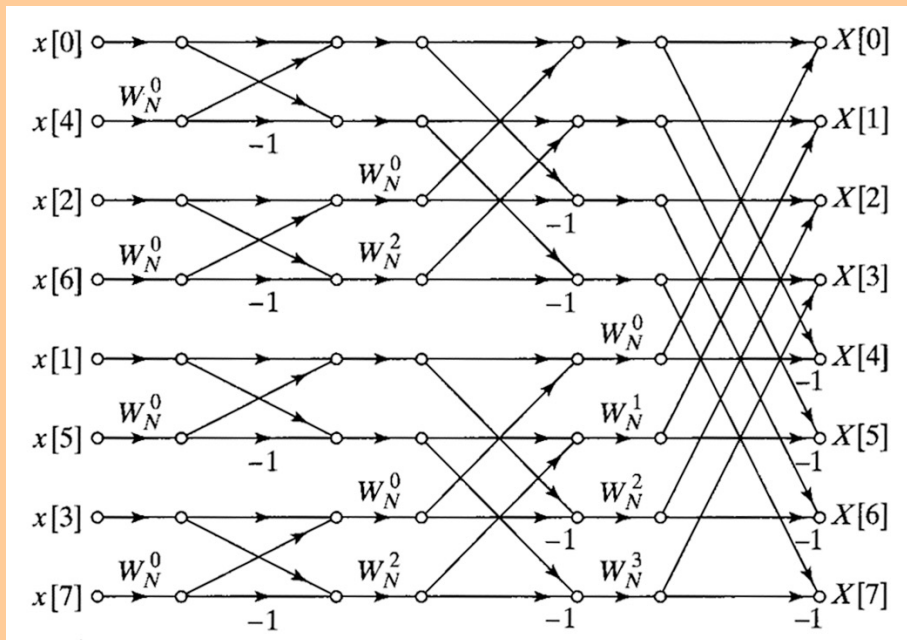
# Alternate DIT FFT structures (continued)

- DIT structure with input bit-reversed, output natural (OSYP 9.11):

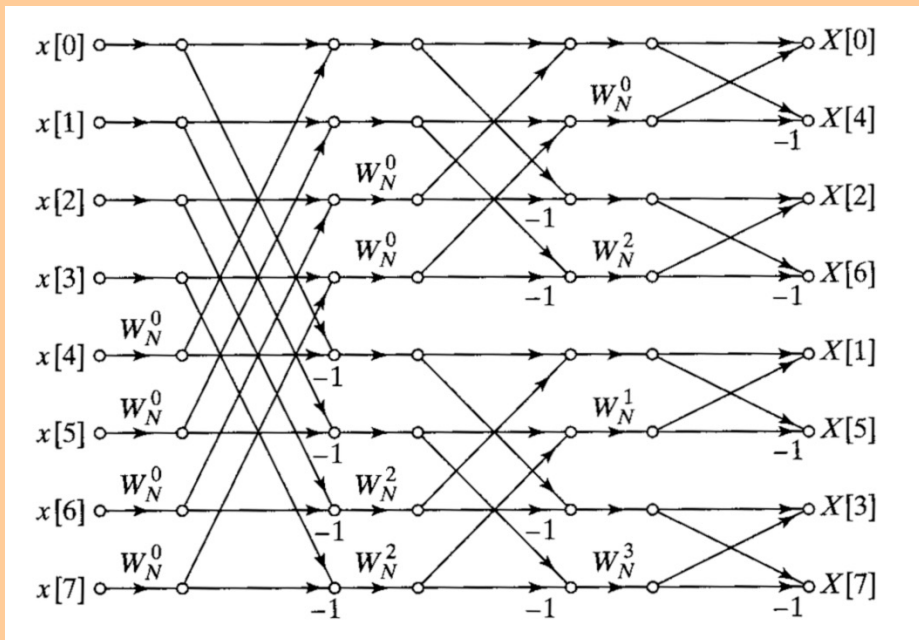


# Alternate DIT FFT structures (continued)

## ■ The original DIT structure (OSYP 9.11):

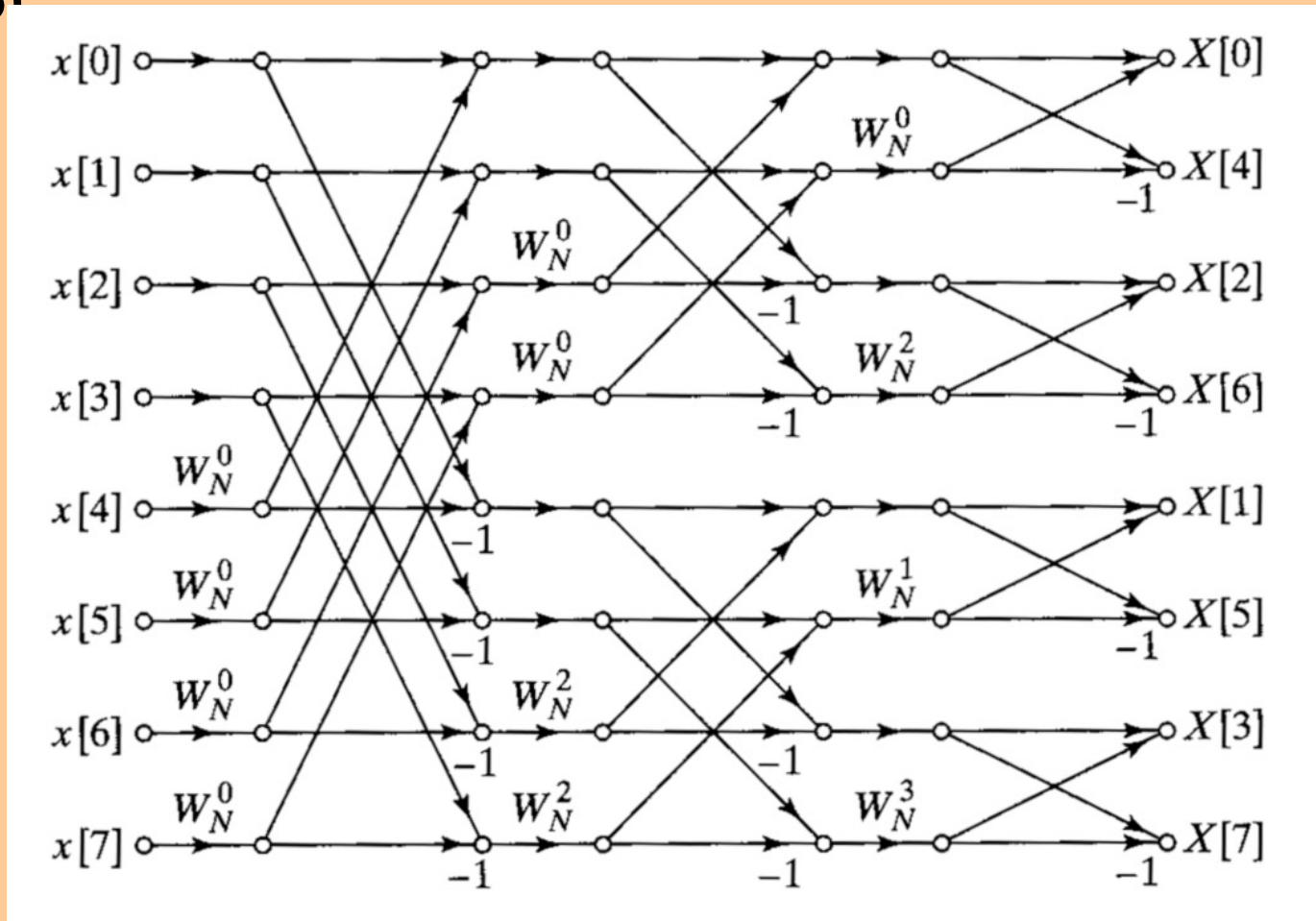


## ■ Rearranged structure (OSYP 9.15):



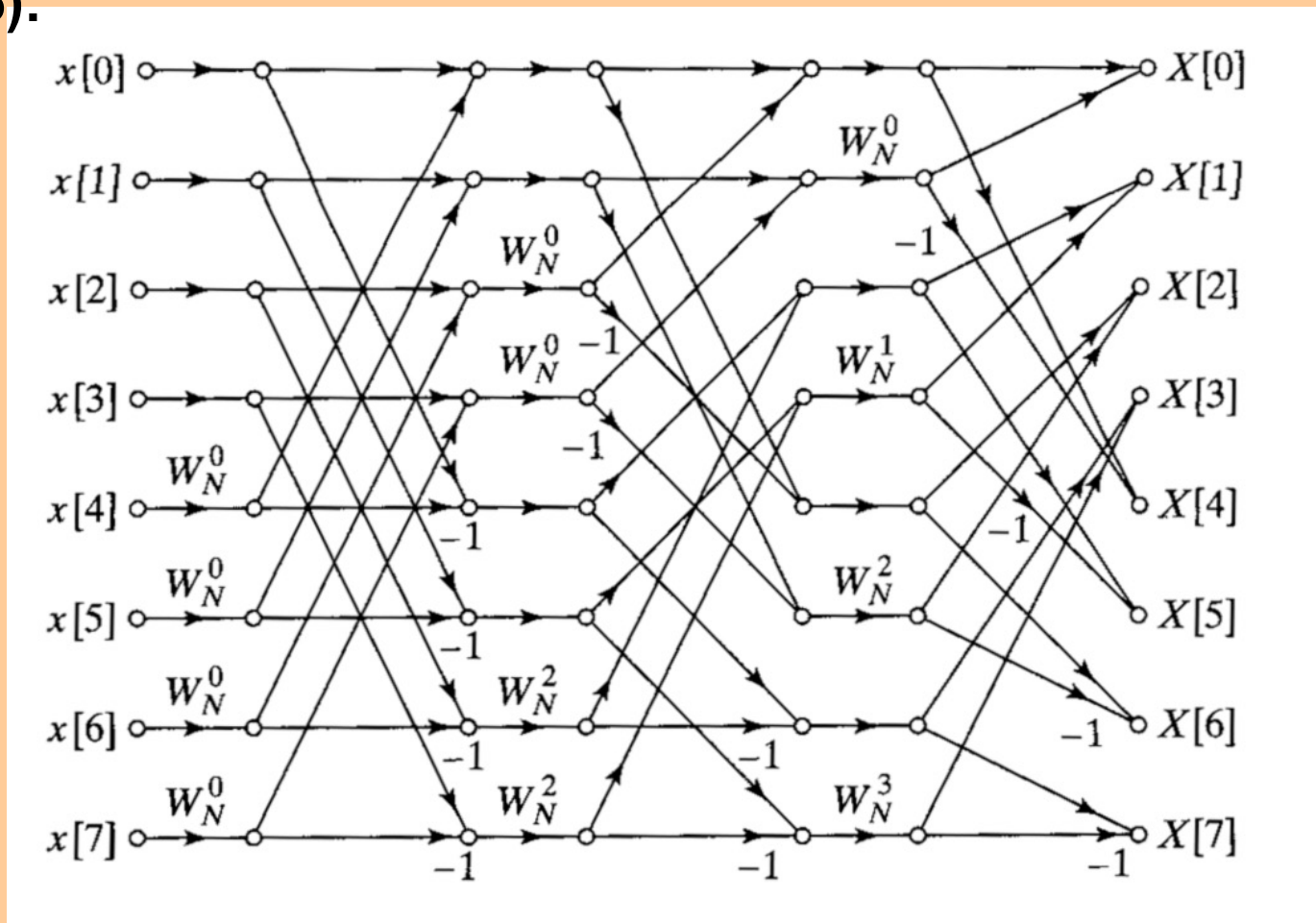
# Alternate DIT FFT structures (continued)

- DIT structure with input natural, output bit-reversed (OSYP 9.15)



# Alternate DIT FFT structures (continued)

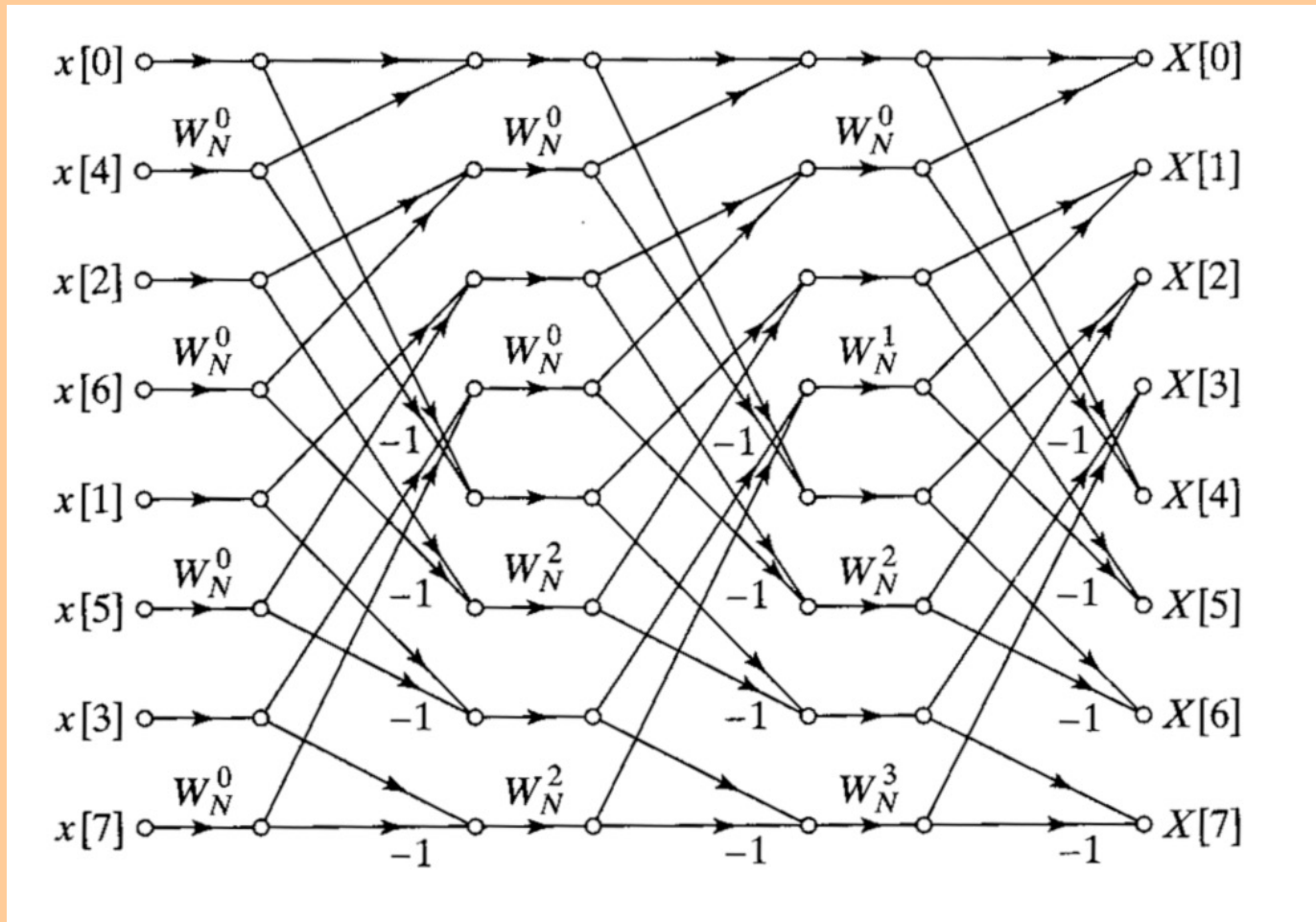
- DIT structure with both input and output in natural order (OSYP 9.16):





## Alternate DIT FFT structures (continued)

- **DIT structure with same structure for each stage (OSYP 9.17):**





# Comments on alternate FFT structures

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- **A method to avoid bit-reversal in filtering operations is:**
  - Compute forward transform using natural input, bit-reversed output (as in OSB 9.10)
  - Multiply DFT coefficients of input and filter response (both in bit-reversed order)
  - Compute inverse transform of product using bit-reversed input and natural output (as in OSB 9/14)
- **Latter two topologies (as in OSYP 9.16 and 9.17) are now rarely used**

# The decimation-in-frequency (DIF) FFT algorithm

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- **Introduction: Decimation in frequency is an alternate way of developing the FFT algorithm**
- **It is different from decimation in time in its development, although it leads to a very similar structure**

# The decimation in frequency FFT (continued)

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- Consider the original DFT equation ....

$$X[k] = \sum_{n=0}^{N-1} x[n] W_N^{nk}$$

- Separate the first half and the second half of time samples:

$$\begin{aligned} X[k] &= \sum_{n=0}^{(N/2)-1} x[n] W_N^{nk} + \sum_{n=N/2}^{N-1} x[n] W_N^{nk} \\ &= \sum_{n=0}^{(N/2)-1} x[n] W_N^{nk} + W_N^{(N/2)k} \sum_{n=0}^{(N/2)-1} x[n + (N/2)] W_N^{nk} \\ &= \sum_{n=0}^{(N/2)-1} \left[ x[n] + (-1)^k x[n + (N/2)] \right] W_N^{nk} \end{aligned}$$

- Note that these are **not**  $N/2$ -point DFTs

## Continuing with decimation in frequency ...

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$$X[k] = \sum_{n=0}^{(N/2)-1} \left[ x[n] + (-1)^k x[n + (N/2)] \right] W_N^{nk}$$

■ **For  $k$  even, let  $k = 2r$**

$$X[k] = \sum_{n=0}^{(N/2)-1} \left[ x[n] + (-1)^{2r} x[n + (N/2)] \right] W_N^{n2r} = \sum_{n=0}^{(N/2)-1} \left[ x[n] + x[n + (N/2)] \right] W_{N/2}^{nr}$$

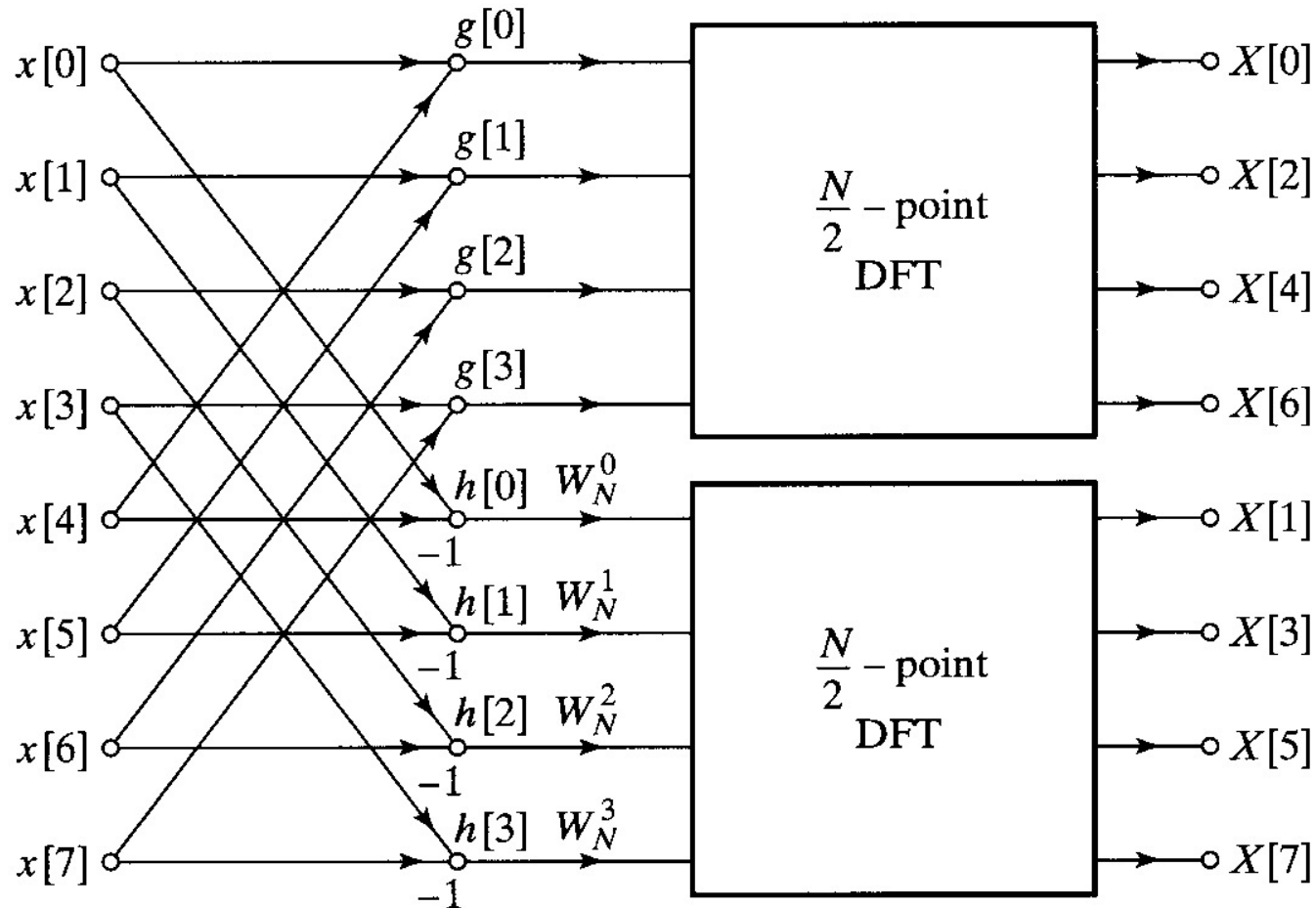
■ **For  $k$  odd, let  $k = 2r + 1$**

$$\begin{aligned} X[k] &= \sum_{n=0}^{(N/2)-1} \left[ x[n] + (-1)^{2r} (-1) x[n + (N/2)] \right] W_N^{n(2r+1)} \\ &= \sum_{n=0}^{(N/2)-1} \left[ x[n] - x[n + (N/2)] \right] W_N^n W_{N/2}^{nr} \end{aligned}$$

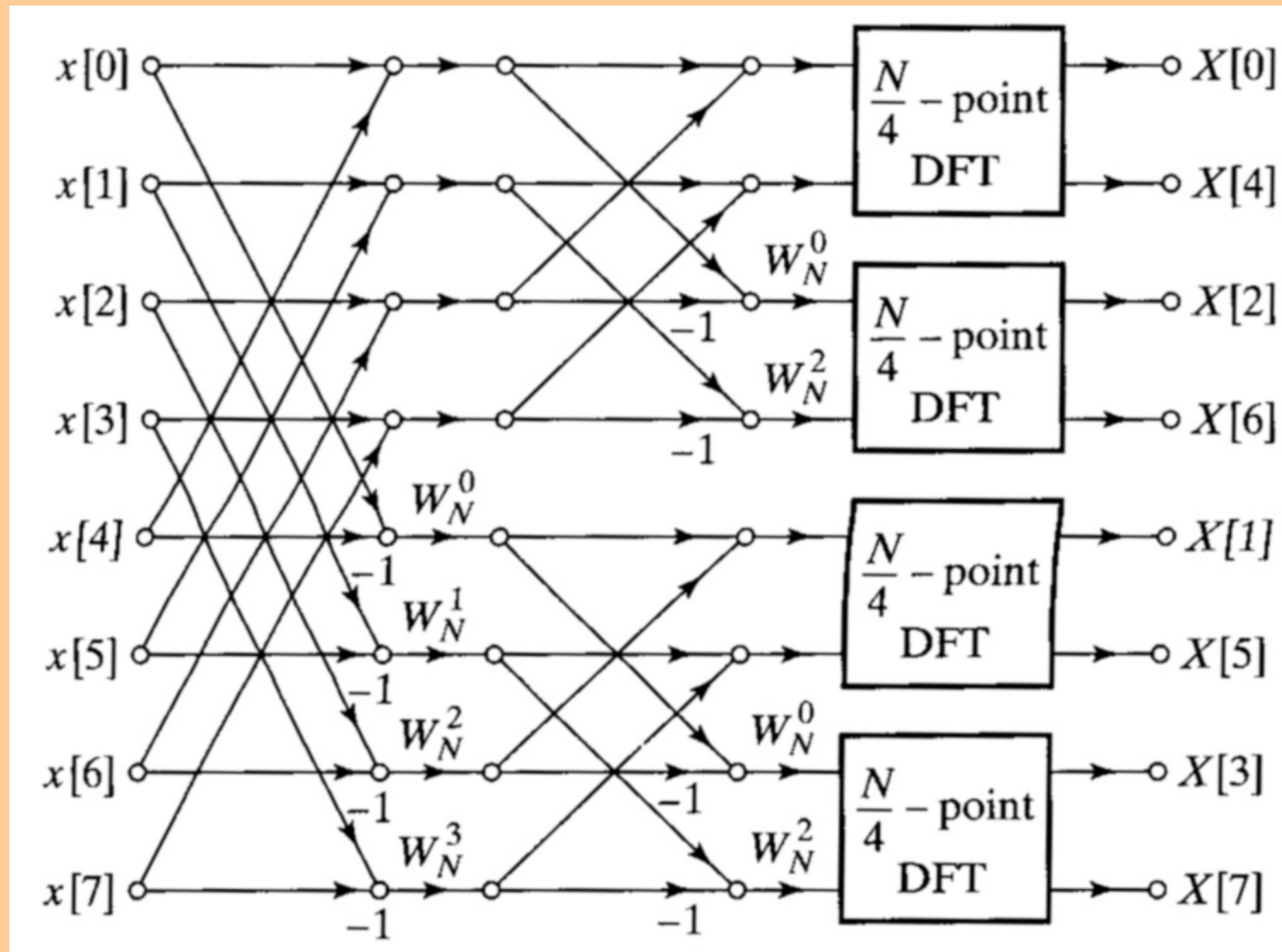
■ **These expressions are the  $N/2$ -point DFTs of**

$x[n] + x[n + (N/2)]$  and  $[x[n] - x[n + (N/2)]] W_N^n$

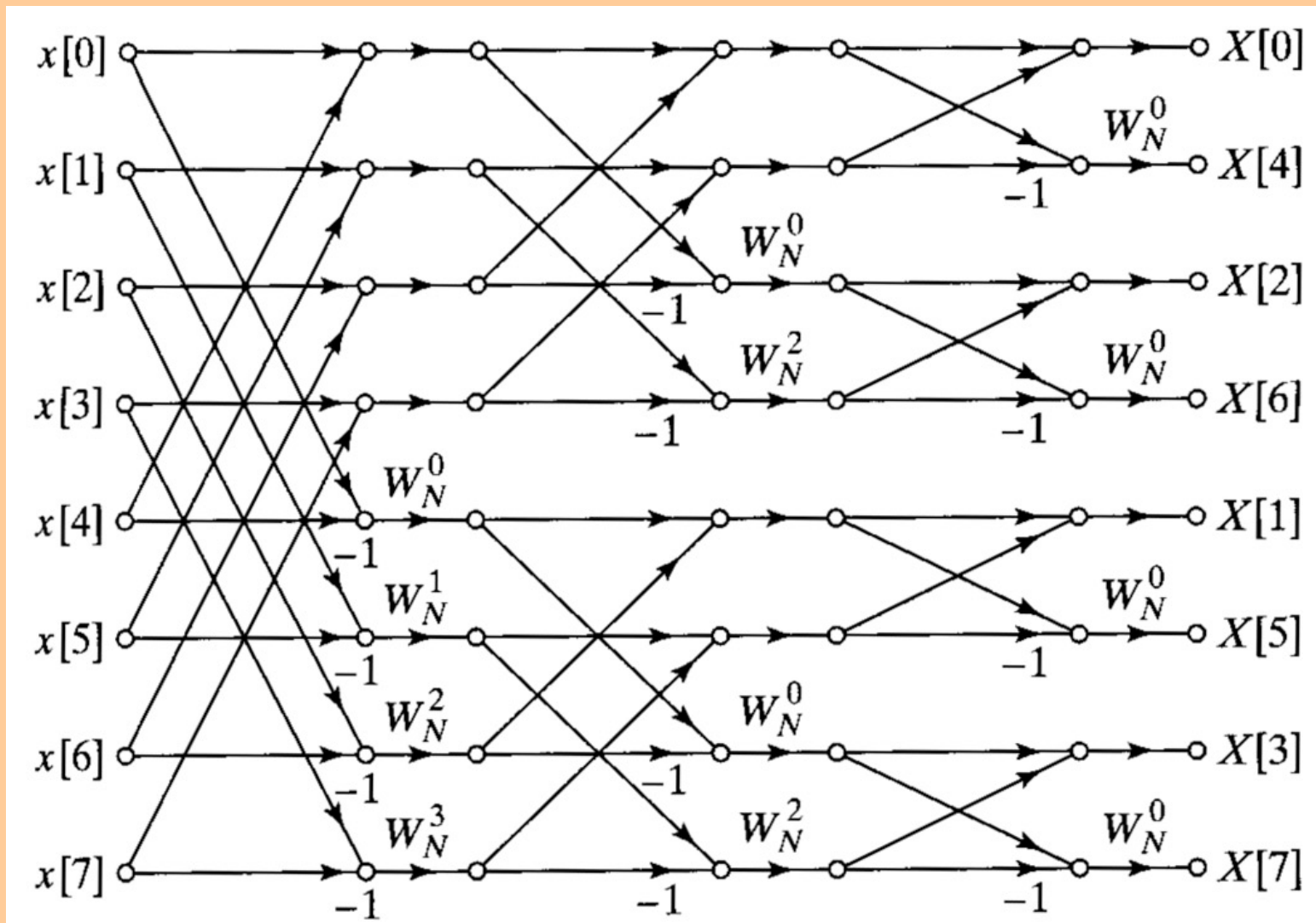
# These equations describe the following structure:



Continuing by decomposing the odd and even *output* points we obtain ...



... and replacing the  $N/4$ -point DFTs by butterflies we obtain



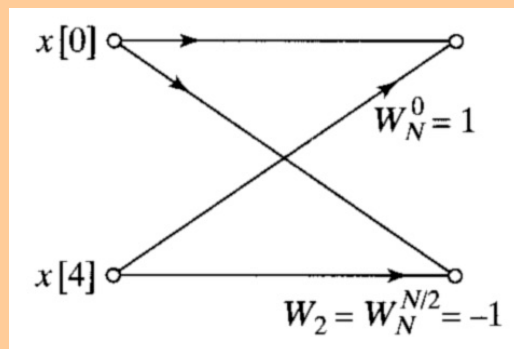


# The DIF FFT is the *transpose* of the DIT FFT

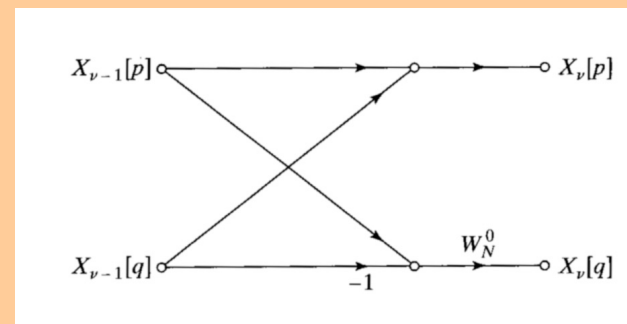
## ■ To obtain flowgraph transposes:

- Reverse direction of flowgraph arrows
- Interchange input(s) and output(s)

## ■ DIT butterfly:



## DIF butterfly:



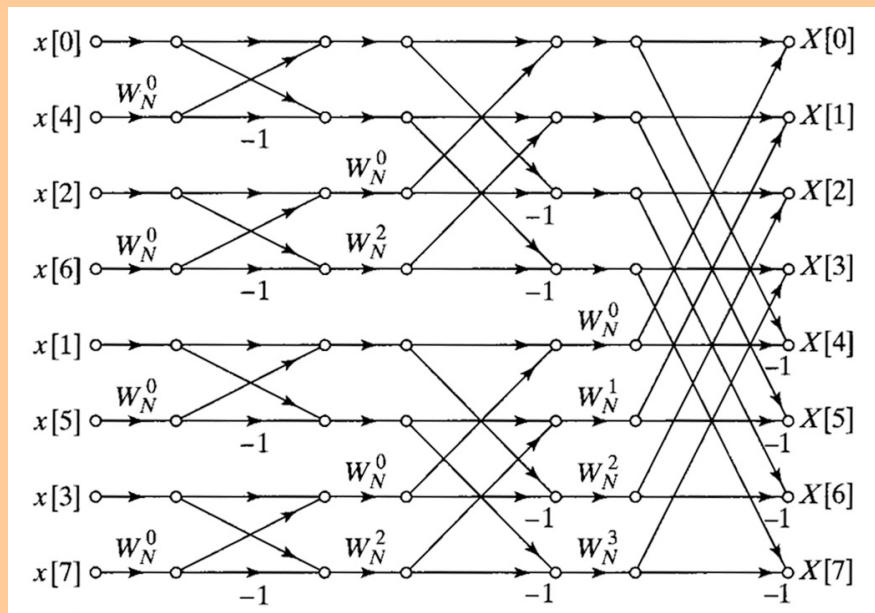
## ■ Comment:

- We will revisit transposed forms again in our discussion of filter implementation

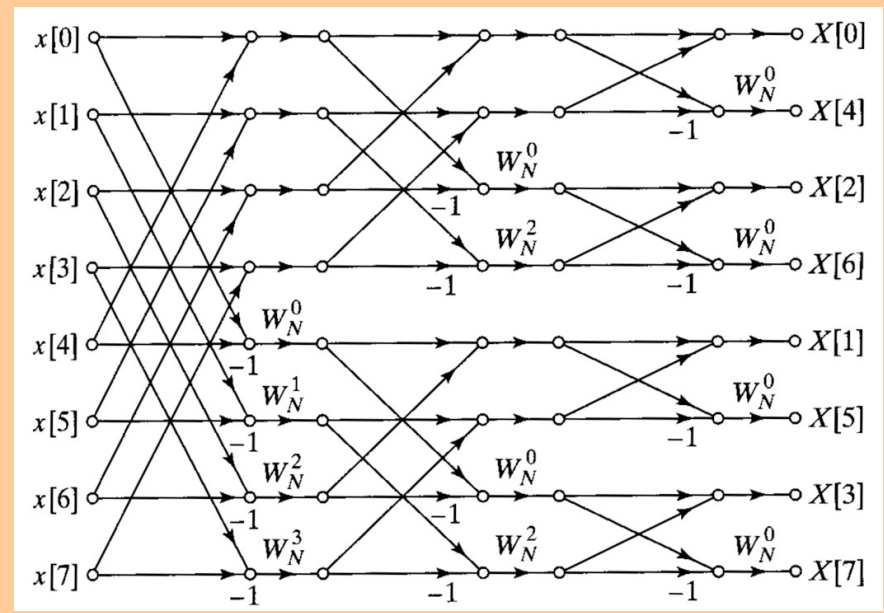
# The DIF FFT is the *transpose* of the DIT FFT

## ■ Comparing DIT and DIF structures:

### DIT FFT structure:



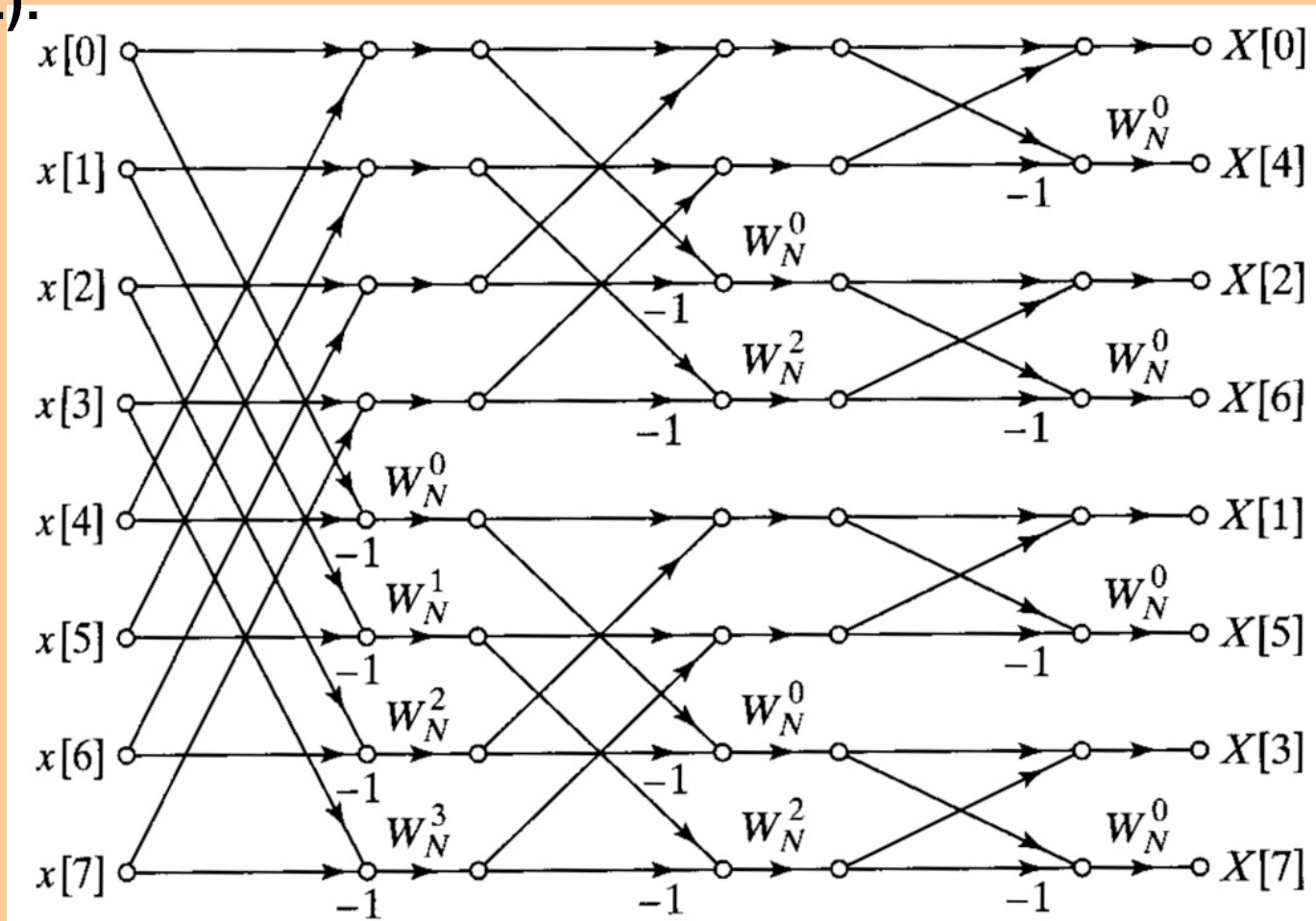
### DIF FFT structure:



## ■ Alternate forms for DIF FFTs are similar to those of DIT FFTs

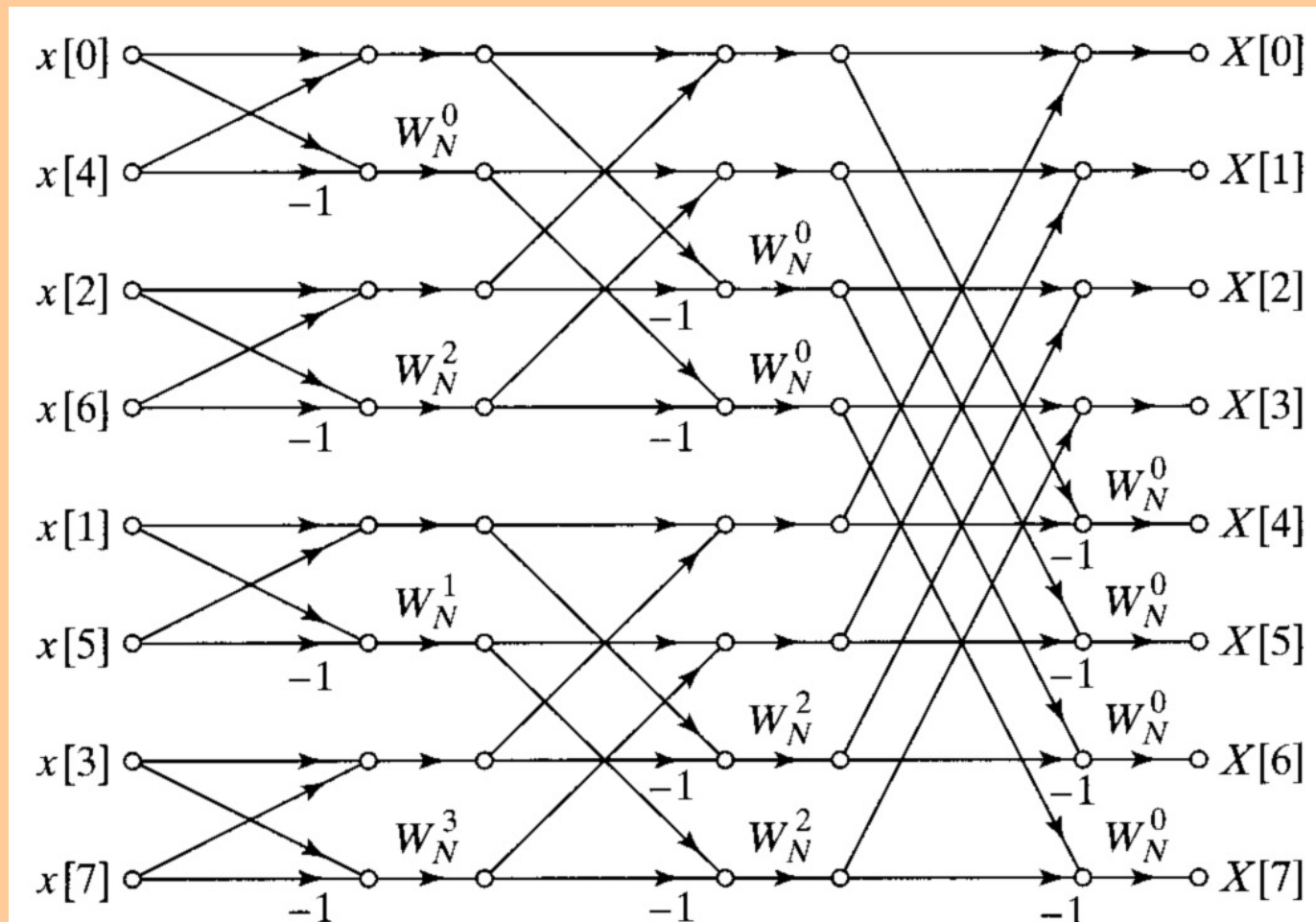
# Alternate DIF FFT structures

- DIF structure with input natural, output bit-reversed (OSYP 9.22):



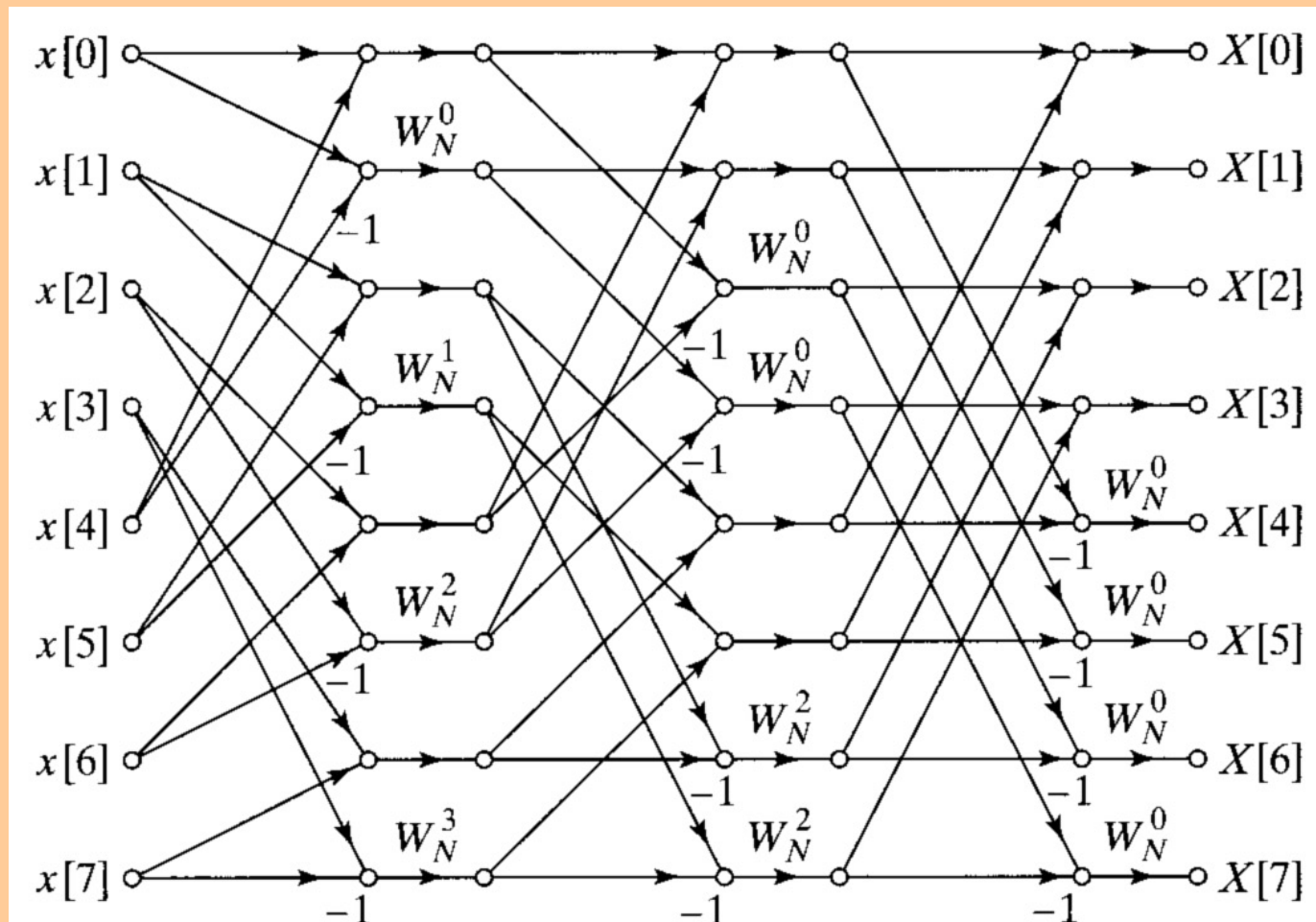
# Alternate DIF FFT structures (continued)

- DIF structure with input bit-reversed, output natural (OSB 9.22):



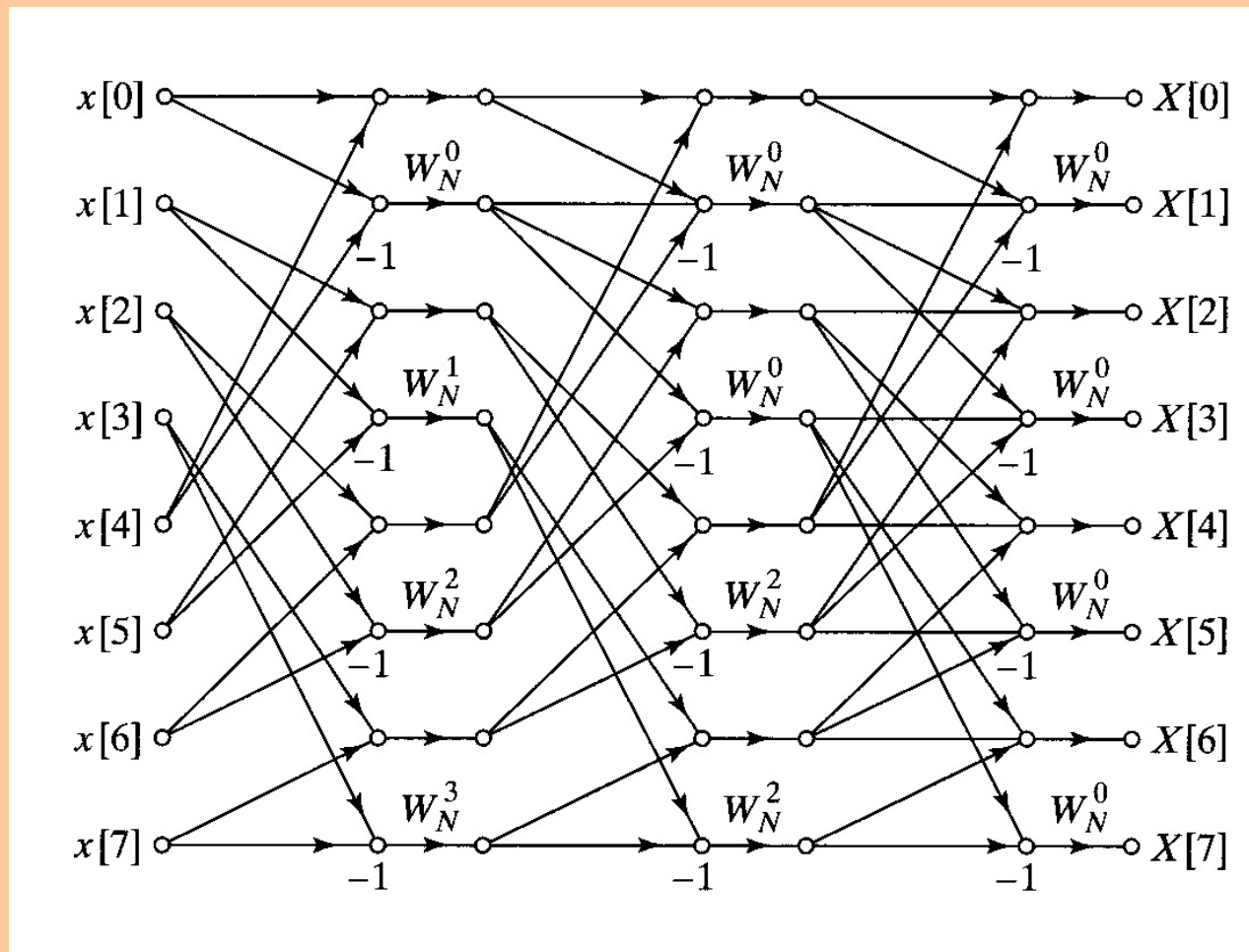
# Alternate DIF FFT structures (continued)

- DIF structure with both input and output natural (OSYP 9.24):



# Alternate DIF FFT structures (continued)

- DIF structure with same structure for each stage (OSYP 9.25):



# FFT structures for other DFT sizes

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- Can we do anything when the DFT size  $N$  is not an integer power of 2 (the non-radix 2 case)?
- Yes! Consider a value of  $N$  that is not a power of 2, but that still is highly factorable ...

Let  $N = p_1 p_2 p_3 p_4 \dots p_v$ ;  $q_1 = N / p_1$ ,  $q_2 = N / p_1 p_2$ , etc.

- Then let

$$\begin{aligned} X[k] &= \sum_{n=0}^{N-1} x[n] W_N^{nk} \\ &= \sum_{r=0}^{q_1-1} x[p_1 r] W_N^{p_1 r k} + \sum_{r=0}^{q_1-1} x[p_1 r + 1] W_N^{(p_1 r + 1)k} + \sum_{r=0}^{q_1-1} x[p_1 r + 2] W_N^{(p_1 r + 2)k} + \dots \end{aligned}$$



## Non-radix 2 FFTs (continued)

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- An arbitrary term of the sum on the previous panel is

$$\begin{aligned} & \sum_{r=0}^{q_1-1} x[p_1 r + l] W_N^{(p_1 r + l)k} \\ &= \sum_{r=0}^{q_1-1} x[p_1 r + l] W_N^{p_1 r k} W_N^{lk} = W_N^{lk} \sum_{r=0}^{q_1-1} x[p_1 r + l] W_{q_1}^{rk} \end{aligned}$$

- This is, of course, a DFT of size  $q_1$  of points spaced by  $p_1$

# Non-radix 2 FFTs (continued)

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- In general, for the first decomposition we use

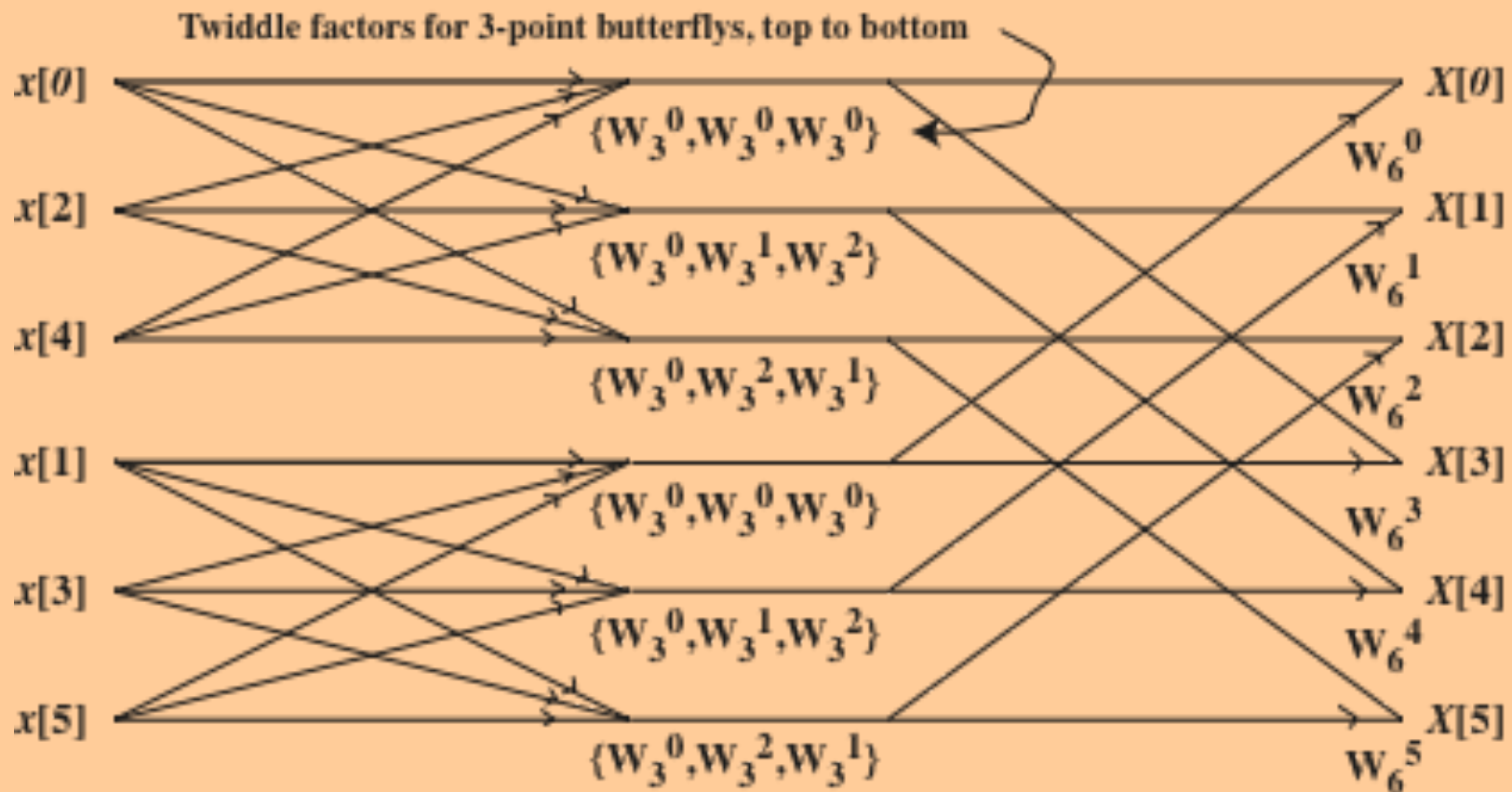
$$X[k] = \sum_{l=0}^{p_1-1} W_N^{lk} \sum_{r=0}^{q_1-1} x[p_1 r + l] W_{q_1}^{rk}$$

- **Comments:**

- This procedure can be repeated for subsequent factors of  $N$
- The amount of computational savings depends on the extent to which  $N$  is “composite”, able to be factored into small integers
- Generally the smallest factors possible used, with the exception of some use of radix-4 and radix-8 FFTs

# An example .... The 6-point DIT FFT

■  $P_1 = 2; P_2 = 3; \quad X[k] = \sum_{l=0}^1 W_6^{lk} \sum_{r=0}^2 x[2r + l] W_3^{rk}$



# Summary

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- **This morning we considered a number of alternative ways of computing the FFT:**
  - Alternate implementation structures
  - The decimation-in-frequency structure
  - FFTs for sizes that are non-integer powers of 2
  - Using standard FFT structures for inverse FFTs
- **Starting on Monday we will begin to discuss digital filter implementation structures**